Abstract—A new voltage-mode triple-input single-output multifunction filter using only two current conveyors is presented. The proposed filter which possesses three inputs and single-output can generate all biquadratic filtering functions at the output terminal by selecting different input signal combinations. The validity of the proposed filter is verified through PSPICE simulations.

Keywords—Active Filters, Voltage mode, Current conveyor

I. INTRODUCTION

Voltage mode circuits using conventional operational amplifiers have the disadvantage of severely reduced bandwidth at higher gains because of the op-amp’s fixed gain-bandwidth product. These disadvantages can be eliminated by using current conveyors [1-3].

Recently, many single and multi-output voltage mode universal biquadratic filters using current conveyors have been reported in the literature [4-10]. Some of these filters employ two positive-type second-generation current conveyors (CCII+) and one voltage follower [5, 9]. Some other filters employ one positive-type second-generation current conveyor, one negative-type second-generation current conveyor (CCII-), two capacitors and three resistors [6, 8]. Liu et al. reported a voltage-mode filter which employs one CCII+, one CCII-, two capacitors and three resistors [8]. Chang proposed a voltage-mode filter which employs three CCII+, two capacitors and three resistors [4]. Later, Chang and Tu presented a voltage-mode filter which uses two CCII+, two capacitors and three resistors with orthogonal control over the natural frequency and the quality factor. However, this filter has four inputs and its transfer function is very complex [5]. Horng proposed a voltage-mode filter which employs three CCII+, two capacitors and two resistors [10].

In this paper, a new voltage-mode triple-input single-output multifunction filter using only two current conveyors is presented. The circuit uses one positive-type, one negative-type second generation current conveyor, three resistors and two capacitors. The proposed filter can generate all second-order filtering functions of low-pass, high-pass, band-pass, notch and all-pass at the output terminal by selecting different input signal combinations. In the literature, there is only one voltage-mode multiple-input, single-output multifunction filter using two current conveyors [5]. However, this circuit has four inputs which results in a complex transfer function. Our proposed filter has only three inputs and has much simpler transfer function. The paper includes the analysis of the non-ideal effects of the proposed filter. It also includes sensitivity analysis and simulations using PSPICE.

II. CIRCUIT DESCRIPTION AND ANALYSIS

The ideal second generation current conveyor shown in Fig. 1 is described by the following relationship [11]:

\[ I_y = 0, \quad V_x = V_y, \quad I_z = \pm I_x \]  

where, \( V_H \) and \( i_H \) denote the voltage and current tracking errors, respectively.

Fig. 1. Block diagram of the second generation current conveyor.

The input impedances for the ideal CCII+ are infinite at terminal \( y \) and zero at terminal \( x \), respectively. The terminal \( z \), that is equivalent to a current generator, possesses infinite output impedance.

Taking the non-idealities of the CCII into account, the above given terminal equation can be rewritten as

\[ I_y = 0, \quad V_x = \beta V_y, \quad I_z = \pm \alpha I_x \]  

where, \( \beta = 1 - \varepsilon_V \) and \( \alpha = 1 - \varepsilon_I \). \( \varepsilon_V \) and \( \varepsilon_I \) denote the voltage and current tracking errors, respectively.
respectively. In Eq. (1) and Eq. (2) the positive sign denotes a CCII+ and the negative sign denotes a CCII-.

The proposed multifunction filter which is illustrated in Fig. 2 employs only one positive-type second generation current conveyor, only one negative-type second generation current conveyor and five passive elements (two capacitors and three resistances). The transfer function of this circuit is given as follows:

\[ V_o = \frac{s^2 R_1 R_2 R_3 C_1 C_2 V_1 + sR_2 R_3 C_1 V_2 + R_1 V_3}{s^2 R_1 R_2 R_3 C_1 C_2 + sR_1 R_2 C_2 + R_1} \]  (3)

The natural frequency and the quality factor of the proposed circuit can be obtained from the denominator polynomial of the transfer functions as follows:

\[ \omega_n = \frac{1}{\sqrt{R_1 R_2 C_1 C_2}} \]  (4)

\[ Q = \sqrt{\frac{C_1}{C_2 R_2 R_3 R_1}} \]  (5)

Taking the non-idealities of the positive-type and negative-type second generation current conveyors given in Eq. (2) into account, the denominator polynomial of the transfer functions for the proposed filter becomes

\[ D(s) = s^2 R_1 R_2 R_3 C_1 C_2 + sR_2 R_3 C_2 + \alpha_1 \alpha_2 \beta_1 \beta_2 R_1 \]  (6)

Using Eq. (6), the natural frequency and the quality factor can be recalculated as

\[ \omega_n = \frac{\alpha_1 \alpha_2 \beta_1 \beta_2}{\sqrt{R_1 R_2 C_1 C_2}} \]  (7)

\[ Q = \sqrt{\frac{\alpha_1 \alpha_2 \beta_1 \beta_2 C_1}{C_2 R_2 R_3 R_1}} \]  (8)

Note that \( \omega_n \) and \( Q \) are adjustable by \( R_1 \) and \( R_2 \) (or \( R_3 \)) and \( BW \) is adjustable by \( R_1 \).

III. SENSITIVITY ANALYSIS

The ideal sensitivities of the natural frequency and the quality factor with respect to passive components are calculated as follows:

\[ S_{R_1}^{\omega} = 0 \]  (9)

\[ S_{R_2}^{\omega} = S_{R_3}^{\omega} = S_{C_1}^{\omega} = S_{C_2}^{\omega} = -\frac{1}{2} \]  (10)

\[ S_{R_1}^{Q} = 1 \]  (11)

\[ S_{C_1}^{Q} = \frac{1}{2} \]  (12)

\[ S_{R_3}^{Q} = S_{R_2}^{Q} = S_{C_2}^{Q} = -\frac{1}{2} \]  (13)

From the above calculations, it can be seen that all sensitivities are fairly small.
From Eq. (7) and (8), the non-ideal sensitivities can be found as follows:

\[ S_{R_1}^\alpha = 0 \tag{15} \]

\[ S_{R_2}^\alpha = S_{C_1}^\alpha = S_{C_2}^\alpha = -\frac{1}{2} \tag{16} \]

\[ S_{\alpha_1}^\alpha = S_{\alpha_2}^\alpha = S_{\beta_1}^\alpha = S_{\beta_2}^\alpha = \frac{1}{2} \tag{17} \]

\[ S_{R_1}^\beta = 1 \tag{18} \]

\[ S_{R_2}^\beta = S_{R_3}^\beta = S_{C_1}^\beta = -\frac{1}{2} \tag{19} \]

\[ S_{C_1}^\beta = \frac{1}{2} \tag{20} \]

\[ S_{\alpha_1}^\beta = S_{\alpha_2}^\beta = S_{\beta_1}^\beta = S_{\beta_2}^\beta = \frac{1}{2} \tag{21} \]

From the above relations, it can be seen that all sensitivities are also fairly small.

IV. SIMULATIONS AND RESULTS

The validity of the filter was tested using PSPICE. Bias currents and voltages of the positive-type and negative-type second generation current conveyor were taken as \( I_1 = I_2 = 200 \, \mu A \), \( V_{cc} = 5 \, V \), and \( V_{ee} = -5 \, V \) respectively.

For these simulations the passive components were chosen as, \( C_1 = C_2 = 1 \, nF \) and \( R_1 = R_2 = 1 \, k\Omega \), \( R_2 = 2 \, k\Omega \).

Using the above parameters, the simulation results for the second-order low-pass, high-pass, band-pass and band-reject filter responses are shown in Fig. 3 and all-pass filter response is shown in Fig. 4, respectively.

![Fig. 3. PSPICE simulation results for the triple-input single-output voltage mode multifunction filter.](image1)

![Fig. 4. PSPICE simulation results for the triple-input single-output voltage mode multifunction filter showing all-pass response.](image2)

V. CONCLUSION

We have presented a new voltage-mode triple-input single-output multifunction filter using only two current conveyors. The validity of the proposed filter has been demonstrated through PSPICE simulations. The proposed filter has the following advantages: (i) Use of only two current conveyors; (ii) Realization of all the basic filter functions: low-pass, high-pass, band-pass, notch and all-pass filter response from same configuration; (iii) Low passive and active sensitivities; (iv) Capability of adjusting \( \omega_o \) and \( Q \) are adjustable through resistors and capacitors; (v) Orthogonal control over \( \omega_o \) and \( Q \).

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